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Characterization of the NEP of Mid-Infrared Upconversion Detectors

Rasmus Lyngbye Pedersen, Lasse Høgstedt, Ajanta Barh, Lichun Meng, and Peter Tidemand-Lichtenberg

Abstract—We present a scheme to estimate the noise equivalent power (NEP) of the frequency upconversion detectors (UCDs), detecting mid-infrared (MIR) light. The NEP of UCD is a combined contribution of NEPs from the upconversion process and from the photodetector, used for detecting the upconverted signal. The 2 – 5 µm MIR range is particularly investigated in this letter using a bulk periodically poled lithium niobate based CW-intracavity UCD. We measured the NEP of UCD as 20 fW/√Hz at MIR wavelength of 3.39 µm. We showed that the limiting factor here is not the noise from the upconversion process (estimated NEP is 2.3 fW/√Hz at 3.39 µm), but from the electrical noise in the photodetector itself. We also compared the performance of our UCD with previously published results and with market available direct MIR detectors. Additionally, we measured the optical noise of the UCD over its working spectral range (2.9 – 3.6 µm) and compared with numerical simulation.

Index Terms—Infrared detectors, Noise, Nonlinear optical devices, Sensor systems and applications

I. INTRODUCTION

FOR a number of applications, spectroscopy in the mid-infrared (MIR: 2 - 10 µm) spectral range is desirable. Several molecules and materials have distinct and strong absorption/emission features in the MIR due to the rotational-vibrational transitions of the molecular bonds. For instance, in the field of combustion research, molecules of particular interest with MIR-features are: sulfate compounds (SO₂, CS₂, H₂S, OCS), hydrogen halides (HCl and HF), small hydrocarbons (CH₄, C₂H₂, C₂H₄, C₂H₆), toxic species like HCN, and hydrocarbon radicals (e.g. CH₃). To detect these substances with high sensitivity, good light sources and detectors are needed. In the last decades MIR sources have seen much improvement with advances made to quantum cascade lasers, optical parametric oscillators and fiber lasers, both for narrow linewidth applications and for broadband-high brightness sources. In contrary, the development of detectors have struggled by the inherent thermal noise in low bandgap materials (PbS, PbSe, HgCdTe), and because there is a significant amount of thermal radiation in the MIR from the environment at room temperature. These limitations have been circumvented by using parametric frequency upconversion to move the MIR signal to the visible or near infrared (NIR) range [1]. This technique has seen rapid development within the last decade.

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II. THEORY

Upconversion is realized by mixing MIR signal photons at frequency \(\omega_{IR}\) with pump photons at frequency \(\omega_p\) to generate VIS/NIR upconverted photons at frequency \(\omega_{up}\), when satisfying the conservation of energy \(\omega_{IR} + \omega_p = \omega_{up}\). This paper is particularly targeting the 2 - 5 µm MIR spectral range for gas species detection, and will focus on periodically poled lithium niobate (PPLN)[8] as the nonlinear mixing crystal. To yield the highest conversion efficiency, the quasi-phase matching (QPM) condition should be satisfied [9] as follows: \(k_{up} - k_{IR} - k_p - 2\pi/\Lambda = 0\), where, \(k_{up}\), \(k_{IR}\), and \(k_p\) are the wavevectors of the upconverted (up) light, MIR (IR) signal, and the pump (p), respectively, and \(\Lambda\) is the poling period of the PPLN. To characterize the NEP of an upconversion detector, it is important to consider both the upconversion quantum efficiency (\(\eta_{up}\)) [10] and the optical noise generated in the upconversion module. In the following we briefly discuss the sources of optical noises and give a recipe to calculate the NEP.

A. Optical Noise Originating from the Upconversion

To our best knowledge, the two primary noise sources in a PPLN upconversion module are upconverted thermal [4] and spontaneous parametric downconversion (SPDC) noise [11]. According to Kirchhoff's radiation law, the PPLN generates thermal radiation in the 2 - 5 µm range, emitted in all direction, due to its finite temperature and absorption coefficient. Similarly, the pedestal effect caused by the random duty cycle (RDC) error can enhance the non-phase-matched process [11], which results in generation of broadband SPDC photons, see Fig. 1(a). The part of infrared radiation (thermal
and SPDC) that satisfies the phase matching condition for the upconversion process, contributes to the optical noise.

The two noise terms can be treated similarly. \( dP_{IR}^j \) is the infrared radiation power from a slab of thickness \( dz \) inside the PPLN over a solid angle \( d\Omega \), where \( j \) can be thermal \((th)\) or SPDC \((SPDC)\), see Fig. 1(b). The associated upconverted power \( P_{up}^j \) is found by the integral:

\[
P_{up}^j(\lambda_{IR}, p_j) = B_j \int dP_{IR}^j(\lambda_{IR}, p_j) \eta_{up}(L-z, \lambda_{IR}, \Omega) dz d\Omega
\]

(1)

The pre-factor \( B_j \) originates from the upconversion of the thermal radiation/SPDC process, \( p_j \) is parameters of the two processes, where \( p_{th} = T \) (temperature of PPLN) and \( p_{SPDC} = \sigma \) (RDC error). The \( \lambda_{IR} \) is the wavelength of the infrared noise (thermal and SPDC). The total upconverted noise power is \( P_N = P_{th}^j + P_{SPDC}^j \).

**B. Noise Equivalent Power (NEP)**

Based on the discussion in the previous section, the average power of the background noise \( P_N \) due to upconverted SPDC and thermal radiation can be calculated theoretically. The performance of the photodetector, used to detect the upconverted signal, should also be considered when evaluating the performance of the upconversion detector (upconversion module + photodetector). The parameter, NEP, is widely used to characterize the sensitivity of detectors. In this section, we calculate the NEP of the upconversion detector theoretically. Assuming a photodetector is used for the detection of the upconverted signal, the SNR of the upconversion detector is

\[
\text{SNR} = \frac{i_s}{\sqrt{\sigma_N + \sigma_{det}^2}},
\]

(2)

where \( i_s = P_{IR}\eta_{up}\eta_{det}q/\hbar \omega_{IR} \) is the current driven by the input signal to the detector, \( q \) is the electron charge. \( \sigma_N = \sqrt{2qBM^2F(i_s + i_n)} \) is the shot noise driven by the current, \( F \) is the excess noise factor of the photodetector, \( M \) is the gain\(^\text{[14]} \), and \( i_n = P_N\eta_{det}q/\hbar \omega_{up} \) is the current driven by the optical noise. Note, the noise photon is assumed to have the same angular frequency as the unconverted signal. \( \sigma_{det} \) is the noise of the photodetector, and it can be approximated as \( NEP_{det}\eta_{det}\sqrt{M}/\hbar \omega_{up} \), where \( NEP_{det} \) is the NEP of the photodetector. Thus, the NEP of the upconversion detector can be calculated by equating \( \text{SNR} = 1 \), which gives

\[
\text{NEP} = \frac{\omega_{IR}\eta_{up}}{\omega_{up}\eta_{up}} \sqrt{\frac{2FM^2P_N\hbar \omega_{up}}{\eta_{det}} + \text{NEP}_{det}^2}.
\]

(3)

It is necessary to emphasize that the shot noise introduced by \( i_s \) is neglected here by assuming it is smaller than the other noise sources. Based on equation 3, the noise source can be separated into two parts: upconversion noise and the photodetector noise. A new parameter can be defined as \( P_{critical} = NEP_{det}^2 M^2 F \omega_{up} \). When \( P_N > P_{critical} \), the performance of the system can be improved using noise reduction strategies, e.g., narrow spectral filtering, spatial filtering, and choosing proper poling period. In contrast, when \( P_N < P_{critical} \), the photodetector becomes the bottleneck of the upconversion detector, and the performance of the upconversion detector cannot be improved significantly by simply reducing the \( P_N \).

In practice, the NEP can be calculated based on the output of the photodetector directly \(^\text{[15]} \):

\[
\text{NEP} = \frac{\sigma_b}{p_r\sqrt{B}},
\]

(4)

where \( \sigma_b \) is the standard deviation of the output of the photodetector and \( p_r \) is the power response of the detector, which can be calculated as \( p_r = (\bar{x}_s - \bar{x}_b)/P_{opt} \), where \( P_{opt} \) is the optical input power of the upconversion detector, and \( \bar{x}_s \) and \( \bar{x}_b \) are the average outputs of the photodetector with and without input signal, respectively.

**III. EXPERIMENTAL SETUP**

Fig. 3 shows the diagram of the experimental setup. A 4 mW continuous wave HeNe laser with a wavelength of 3.39 \( \mu \)m
Fig. 3: Mirrors M1-M7 are HR coated for 1064. M1-M5 are transmissive for 808 nm and M6 and M7 are transmissive for the upconverted and the IR signal, respectively.

was used as the MIR signal. The HeNe beam is guided by M8 and M9 and then focused by L1 (f = 75 mm, germanium lens) into the PPLN crystal forming a beam waist of 80 µm, to match the 1064 nm pump beam waist (90 µm). The HeNe beam is aligned to optimize the overlap between the two beams inside the PPLN. The PPLN was 25 mm long with 10 different poling period channels. The cavity of the 1064 nm laser consists of mirrors M1-M7. The laser crystal is an Nd:YVO4, pumped by a 4 W, 808 nm laser diode (LD). The laser cavity is built with several baffles and mirrors (M2-M5), to remove the remainder of the LD light. The PPLN crystal is between mirrors M6 and M7, and its temperature is controlled by an oven. The upconverted signal is collimated with the lens L2 (f = 60 mm). A pinhole, 700 µm in diameter, is used to remove the non-collinear upconverted thermal radiation. The filter set shown in Figure 3 is used to remove the 1064 nm laser leaking from the cavity, and the 532 nm second harmonic generated in the PPLN. A Thorlabs PDF10A photodetector is used to measure the upconverted signal. To avoid saturation of the detector, ND filters are added in front of L1 to reduce the signal. The photodetector is connected to a Tektronix TDS 102 oscilloscope.

IV. RESULTS AND DISCUSSION

Before determining the optical noise contribution from the upconversion module, the electronic noise contributions from the oscilloscope (osc.) and the photodetector (PD) is characterized by measuring the readout of the osc. over 1 second, with 1 ms time resolution for three cases, see Table I. The noise contribution from the upconversion module is negligible compared with the noise from the PD.

TABLE I: Electric noise*

<table>
<thead>
<tr>
<th>Noise source</th>
<th>Osc.</th>
<th>Osc. and PD</th>
<th>Osc., PD and Upc.</th>
</tr>
</thead>
<tbody>
<tr>
<td>Mean (mV)</td>
<td>0.2</td>
<td>19</td>
<td>85</td>
</tr>
<tr>
<td>STD (mV)</td>
<td>0.26</td>
<td>8.5</td>
<td>8.6</td>
</tr>
</tbody>
</table>

* In this measurement, the PPLN poling period is 22.7 µm and the crystal is at 17 °C.

To determine the power response, a known power input of 0.25 pW was generated by reducing the 4.0 mW HeNe laser beam with ND filters calibrated to ND10.2. The signal generated with this input was 34.4 mV, corresponding to a power response of 1.35 × 10^11 V/W. Combining the power response with the background STD in Eq. 4 gives a NEP of 20 fW/√Hz. If the noise contribution from the upconversion module is negligible, it can be seen from Eq. 3 that the NEP is the NEP of the detector scaled by the conversion efficiency of the upconverter. While this is the case, increasing the conversion efficiency will lower the NEP, while simultaneously increasing the generated optical noise, until the detector is no longer the limiting factor.

The upconversion detector is capable of phase-matching over a wide wavelength range, and the noise power depends on the phase-match condition. In Fig. 4 the measured noise power is shown for different poling periods and compared with the noise power calculated using Eq. 1. There are a few outliers, that do not follow the trend predicted by the calculations. This could be a result of imperfect alignment, as the phase-match could only be optimized at 3.39 µm, before translating the PPLN to a different poling period channel. Alternatively, it might be an effect of different errors in the poling period of each channel. The measured noise power are on average five times larger than the calculated values, but follows the trend suggested by the calculations. The measured noise power depends on the poling period and crystal temperature, which suggests that the source of the additional noise is not stray light from the LD pump. It is at this point unclear what the source of the additional noise is.

As discussed above, the electronic noise in the detector dominates the overall noise. Determining the noise limit set by the noise generated in the upconversion module alone would make it possible to choose a detector that does not further limit detection sensitivity. Assuming the background photon noise is Poisson distributed, an estimated NEP reachable with a better detector can be calculated. For a Poisson distribution, the STD is the square root of the mean number of events. Assuming photon noise level of 500 fW as a rough value, or 2.0 × 10^6 photons at 810 nm, the STD is σ = 1.4 × 10^3 photons.
per second. For an integration time of 1 s, the NEP can be calculated by:

$$\text{NEP} = \frac{\sigma E_p \omega_{IR}}{\eta_{up} \Delta f} = 2.3 \text{fW/\sqrt{Hz}} \quad (5)$$

where $E_p$ is the energy per photon, $\eta_{up} = 6.0\%$ and $\Delta f$ is 0.5 Hz. For the contribution of the detector noise to negligible, $NEP_{det}$ should be much smaller than the value above. If the detector noise is the dominant noise source, the NEP of the upconversion detector can be estimated as $\omega_{IR} / (\omega_{up} \eta_{up}) NEP_{det}$.

V. COMPARISON WITH OTHER MIR DETECTORS

In Table II the NEP determined in this article is compared with two MIR detectors, and previous results using upconversion. The MIR detector examples chosen are market available cutting edge direct MIR detectors, a Vigo photodiode as an example of MIR detector with cryogenic cooling (−78°C), and an InSb photodiode from Teledyne as an example of a liquid nitrogen cooled detector. The combination of the PDF10A photodiode and upconversion (working temperature $\approx 20^\circ \text{C}$) compare favorably with direct detection in terms of NEP. When comparing with the work by Zhou et al.[17], the pure value for the reported NEP is almost three orders of magnitude better than the figure presented here. However, the upconversion pump in the system described in [17] is pulsed, and the NEP and conversion efficiency calculated only for the duration of the pulse. Their upconversion pump laser pulse duration is 5.3 ps, at 17.9 MHz repetition rate, giving a duty cycle of only 0.01 %. With a similar time gating, our noise would decrease four orders of magnitude. When comparing with previous CW upconversion results from [16], it can be seen that the NEP has been improved by almost 3 order of magnitude, while using a detector with a higher NEP. This is a result of the higher $\eta_{up}$ reached here.

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